

# SCVs: A Family Album

## Part 3: The Rectangular Division

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So far, we have looked at the delta loop SCVs. To review, the SCV antennas are the class of self-contained, vertically polarized,  $1-\lambda$  wire antennas and include deltas, quad loops, and open-ended arrays such as the half-square and bobtail curtain. Basic designs require about one wavelength of wire, but doubled versions also work very well. These antennas only become SCVs when their feed points are located so as to maximize vertically polarized radiation and minimize horizontally polarized radiation. Under those conditions, they have less raw gain than when fed for horizontally polarized radiation, but they exhibit very low angle bi-directional patterns broadside to the array with very little response to high-angle radiation. Hence, they are favored by many DXers who will trade gain for a better signal-to-noise ratio for distant signals. Moreover, they require no ground plane.

For 160 meters, where everything is big, the SCV with the smallest vertical and horizontal dimensions is the side-fed rectangle. Squares are much taller with less gain, while half-squares are also taller with only a small margin of extra gain over the rectangle. Deltas are also taller, and longer corner-to-corner than the rectangle. Hence, for most installations requiring an SCV, the rectangle should be the antenna of choice.

The side-fed rectangle was popularized in recent times by K5RP, who called it a magnetic slot in his *Antenna Compendium*, Vol. 2 article.<sup>1</sup> One may get a grip on the side-fed rectangle more easily by deriving it from the side-fed square (or quad) loop. See [Figure 1](#). When used for horizontally polarized radiation, the bottom fed loop has been long known to lower its feed point impedance and increase its gain if elongated vertically. K6STI reported some time back in *QST* on using the loop in this configuration to achieve a  $50 \Omega$  impedance.<sup>2</sup> Also, this configuration is the basis for the Hentenna, an elongated loop with good gain and an impedance matching system.

In free space, there is no up or down. Therefore, laying the square loop on its "side" gives us the same antenna with the field at 90 degrees to the bottom-fed version. Likewise, tilting the elongated loop over on its side produces the magnetic slot or single rectangle. When placed over the earth rather than in free space, the antenna produces mostly vertically polarized radiation. Like its brother, the square loop, it needs no ground plane.

Maximum free space gain from the rectangle occurs when the long unfed sides are about 3 times longer than the short fed side and the loop is brought to resonance. This shows clearly in [Figure 2](#), which tracks free space models (using #12 copper wire). This 3:1 figure is normally not very critical, as the gradual slope of the curve demonstrates. The resonant feed point resistance for the 206-ft long by 68-ft high version is just above  $30 \Omega$ , which is a bit low until we introduce proximity to the earth into the picture.

In fact, the ratio of long side to fed-side of the rectangle for maximum free space gain is frequency dependent. Based on free space models of rectangles resonated at frequencies from 1.8 to 146 MHz, the ratio (R) of the long side to the short side for maximum gain (where F is the frequency in MHz) is as follows:

$$R \approx \sqrt{2 \log(100F)}$$

Like the counterpart relationship between the baseline length and the height of deltas, this approximation holds fairly

well into the 2-meter region, which the #12 wire basis for the models becomes an appreciable part of a wavelength.<sup>3</sup> However, the curve approaching maximum gain is shallow, and little is lost from being slightly off the ideal mark.

Although the ratio of long-to-short side increases for maximum gain in an elongated loop with increases in frequency, the feed point impedance follows normal rules: the more extreme the elongation, the lower the feed point impedance at resonance for the resulting loop. At 2 meters, where the ratio approaches 6:1, the feed point impedance is in the vicinity of  $7 \Omega$ .

### The Single Rectangle

[Figure 3](#) sets the dimensions used in this study for the single rectangle. (We shall look at the K5RP double rectangle a bit farther on.) I shall leave it to builder ingenuity as to how

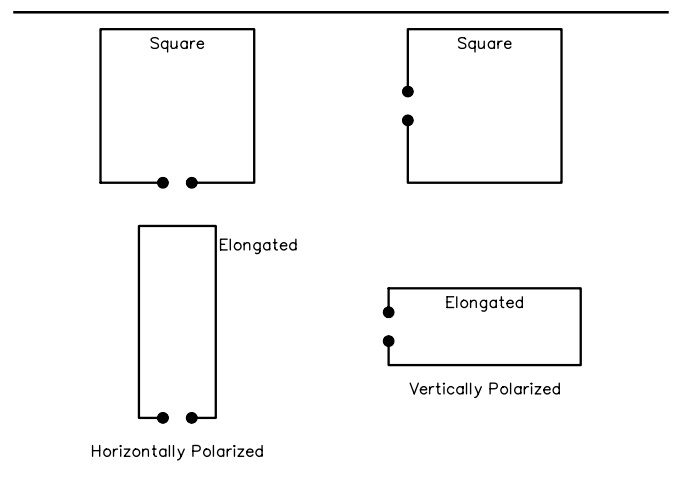


Figure 1—The basis of the rectangular SCV in the quad loop.

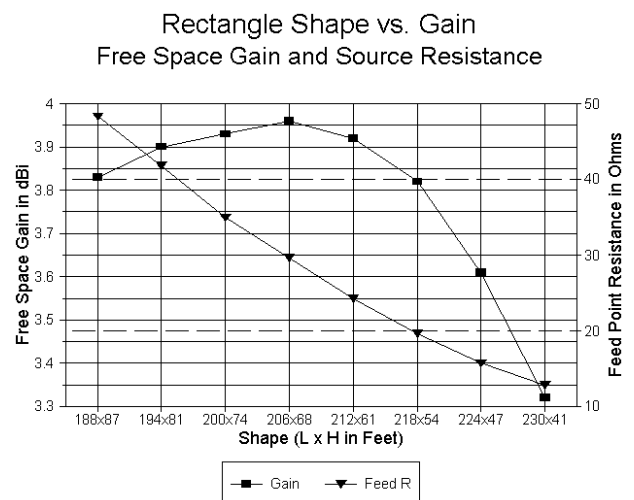
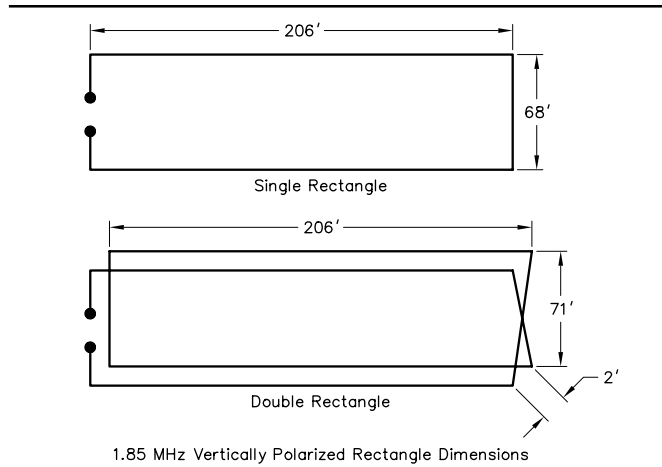


Figure 2—Gain and source resistance of the 160-meter single rectangle in free space.

<sup>1</sup>Notes appear on [page 20](#).

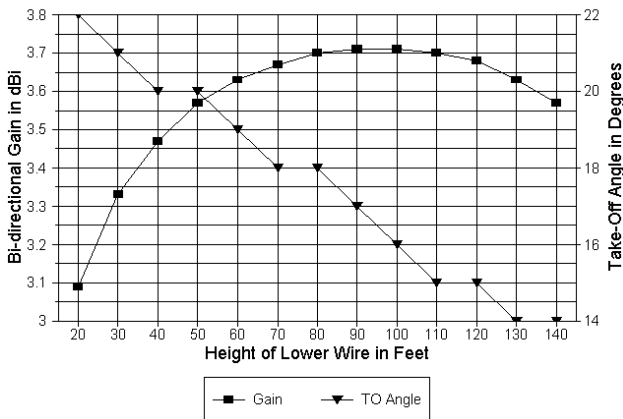
one might hang up this antenna. But before we string a long length of wire, we should ask how high to string it.

**Figure 4** provides modeling data on the single rectangle for bottom wire heights ranging from 20 ft to 140 ft up. The antenna top wire will be 68 ft higher. The left axis records gain over average soil for the various heights. Note that as the bottom wire reaches the 90 to 100-ft mark, the gain levels off and then decreases with further increases in antenna height.

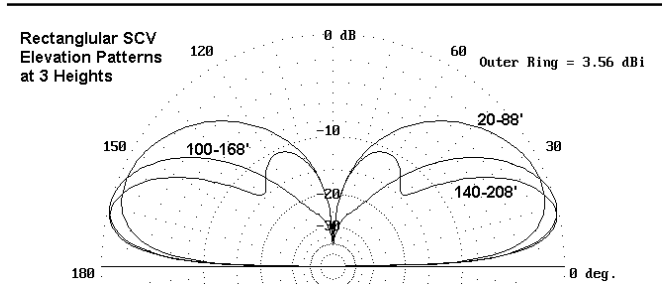


**Figure 3—Dimensions for single and double 160-meter rectangles.**

**Single Rectangle: Gain and TO Angle**  
206x68'; Average Soil; 1.85 MHz



**Figure 4—Single 160-meter rectangle gain and take-off angle at various heights.**



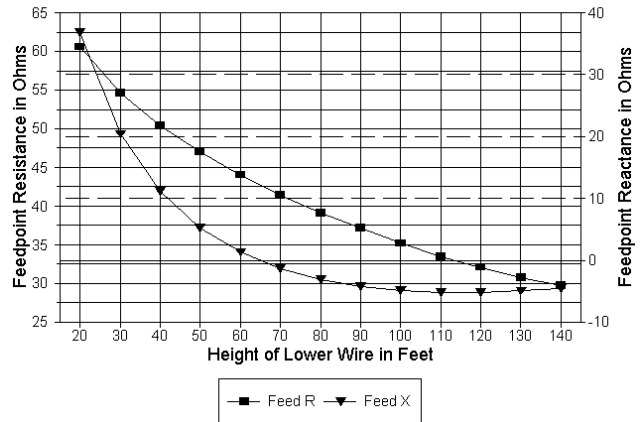
**Figure 5—Three elevation patterns for the single rectangle at different heights.**

The second curve referenced to the right axis records the continuous decrease in the elevation angle of maximum radiation, which runs from 23 degrees for the lowest height surveyed to 14 degrees for the highest.

On the very dubious assumption that we have some choice in how high to place the rectangle, we can also use elevation patterns of the antenna as a guide. **Figure 5** provides three patterns with the bottom wire at 20 ft, 100 ft and 140 ft. The decrease in gain with the 140 ft pattern is already evident, along with the reason for that decrease. Above the height for maximum gain, a secondary lobe begins to appear at a very high angle. With sufficient elevation (above  $\frac{1}{2} \lambda$  or so), the secondary lobe becomes dominant and the overall low angle gain of the antenna begins to decrease dramatically. However, I doubt anyone will ever test this modeling result by placing the bottom wire of the rectangle above 270 ft.

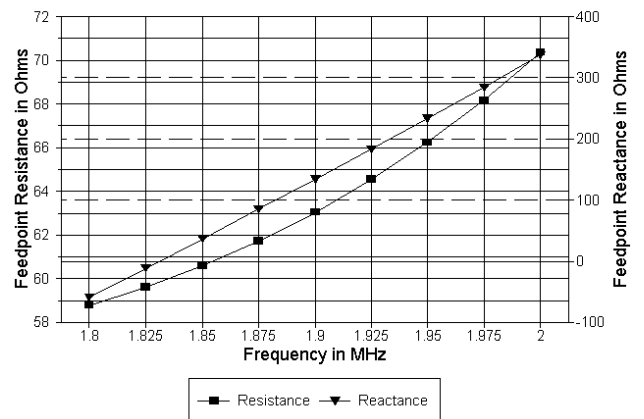
We can see the span of expected feed point impedances in **Figure 6**. At 140 ft, the feed point impedance approaches that of free space models. The resistive component increases in a regular curve as the antenna is lowered to more usable

**Single Rectangle: Feedpoint Impedance**  
206x68'; Average Soil; 1.85 MHz



**Figure 6—Single 160-meter rectangle feed-point impedance at various heights.**

**Single Rectangle: Z vs. Frequency**  
20' Lower Wire; Ave. Soil; 1.8-2.0 MHz



**Figure 7—Single 160-meter rectangle feed-point impedance across the band.**

heights, in accord with reductions in gain as the antenna passes the 90-ft level on its way down. Also notable as an earth-effect is the increase in inductive reactance with lower heights, a fact suggesting that builders of rectangles at low heights might wish to shrink them to bring them to resonance. (However, before adjusting the antenna size, see the notes on matching below.) Notice that the feed point resistance of the antenna at its target frequency (here, 1.85 MHz) is well within the range of direct coaxial cable feed.

As one might expect, the SWR bandwidth of this antenna, like most 160-meter antennas, is fairly narrow. More important is the information in **Figure 7**, a track of the resistance and reactance across the band for the antenna when the lower wire is at 20 ft. Similar data accrue for higher antenna elevations, but with a shift in the resistance curve according to height.

The range of resistance is rather narrow—under  $12\ \Omega$  across the band, as read from the left axis. Moreover, the reactance (as read from the right axis) changes in a very linear fashion. It would be simple enough to enlarge the antenna until it displays inductive reactance across the band. A remotely tuned series capacitor might then compensate for the reactance, leaving a resistive impedance suitable for coaxial cable.

The data developed so far has been over average earth, as it is called. Unfortunately, many of us live over earth significantly poorer than average, while a few lucky souls live on very good earth. **Table 1** provides some guidance as to expectations for various types of soil ranging from very poor (conductivity (C) = 0.001 S/m; dielectric constant (DC) = 5) to very good earth (C = 0.0303; DC = 20). The figures reflect single side-fed rectangles at base heights of 20, 60 and 100 ft. Similar tables might be drawn up for any vertically polarized antenna and show similar differences from one soil type to the next. However, unlike the delta loops, which showed some aberrant progressions with changes in soil conditions, the course of values shows relative smooth curves

**Table 1**  
**160-Meter Single Rectangle Gain, TO Angle, and Feed-point Impedance Over Various Soils, with Different Antenna Heights Above Ground.**

Antenna Height (in feet)	Gain (dBi)	Take-Off Angle (degrees)	Feed-point Impedance (R +/- jX $\Omega$ )
<b>20-88</b>			
Very Poor Soil	0.34	27	65.1 + j30.0
Poor Soil	1.86	25	62.4 + j34.1
Average	3.09	22	60.6 + j36.8
Very Good	5.24	16	55.7 + j37.4
<b>60-128</b>			
Very Poor Soil	1.09	23	42.5 - j 1.7
Poor Soil	2.55	21	43.1 + j 0.2
Average	3.63	19	44.0 + j 1.3
Very Good	5.78	14	44.2 + j 3.1
<b>100-168</b>			
Very Poor Soil	1.24	20	33.3 - j 5.1
Poor Soil	2.67	18	34.3 - j 4.8
Average	3.71	16	35.2 - j 4.9
Very Good	6.16	13	36.1 - j 4.4
<b>Soil types</b>	<b>Conductivity (S/m)</b>	<b>Dielectric Constant</b>	
Very Poor Soil	0.001	5	
Poor Soil	0.002	13	
Average	0.005	13	
Very Good	0.0303	20	

from one soil type to the next for rectangles.

The tables are based on uniform soil in every direction from the antenna for distance great enough to fully affect the far field. Of great interest is the lack of significant change in the antenna feed point impedance with changes in soil type. It is dubious whether one can effect significant performance improvements in this or any other SCV by doctoring the soil in the immediate vicinity of the antenna. On the other hand, the soil at a distance of  $2\ \lambda$  and more from the antenna is usually beyond control.

The better the distant soil, the better the low angle radiation from the antenna. **Figure 8** shows the contrast between very poor and very good soil. Equally important with the gain improvement is the lowered angle of maximum radiation. Were the curves graphically equalized, the higher-angle response would be very little different. However, the low-angle response from the antenna over very good soil is very much enhanced. Intermediate soil types provide intermediate curve shapes.

### The Double Rectangle

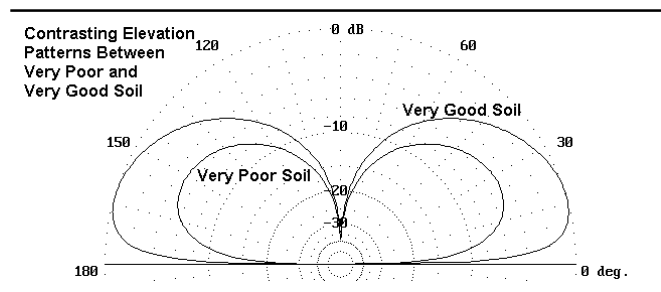
For various reasons, some antenna builders prefer to feed their antennas with parallel transmission line and use an ATU to effect a match. One effective way to do this is to use the antenna as an impedance transformer by doubling it and making a mobius-strip crossing at the end opposite the feed point. This is the double rectangle shown in **Figure 3**, and developed by K5RP.

The sketch shows two significant features. First, the loop must be slightly fatter vertically for maximum gain relative to the single loop. Second, the spacing between the loops has very little effect on antenna gain. In fact, models of the double loop with a space that ranges from 1-ft up to 12 ft showed only a 0.01 dB difference in gain.

What did change with changes in spacing was the required total loop size for resonance. The closer the wires, the larger the loop size. Part of this size increase stems from the fact that the crossing wires are longer for wider spacing, thus occasioning smaller outside dimensions. However, the other part is a function of minor interactions between the loops. With a constant length of 206 ft for the double rectangle array, the loop height was 70 ft with a 12-ft spacing and 71-ft for a 2-ft spacing.

The remainder of the data was generated on the basis of a 2-ft space between the wires, simply because that is most likely a more convenient construction distance. Spacers can be almost anything that insulates and is light weight. It is essential that the crossing wires at the far end of the array be well insulated from each other, although spacing appears not to be critical.

The double rectangle exhibits almost a half-dB of additional gain beyond that of the single rectangle. **Figure 9** shows the gain and take-off angle for lower-wire heights from 20 ft to



**Figure 8—Contrasting elevation patterns of the rectangle over very poor and very good soil.**

140 ft, which permits direct comparison with the corresponding chart for the single rectangle. The curves are highly congruent, with peak gain at about 100 ft (with the top wire at 171 ft). Since the resolution of take-off angles is one degree, the “stair-step” form of that curve should be no surprise, and 1-degree differences between the graphs for the single and double rectangles are meaningless.

The “transformer” action of the double rectangle configuration appears clearly in **Figure 10**. The forms of both the resistance and the reactance curves are almost identical to those of the single rectangle, but the double rectangle values are almost exactly 4 times those for the single rectangle. Both resistance and reactance are multiplied.

The curve suggests that almost any parallel feed line might be used to feed the double rectangle. A link-coupled tuner should be able to handle the range of resistance and reactance across 160 meters. In fact, the more typical network tuner with its 4:1 output balun should not be heavily challenged by the impedances presented by the double rectangle.

Like the single rectangle, soil quality affects the far field pattern fairly strongly without affecting the feed-point impedance very much. **Table 2** presents data for the double rectangle at the same baseline heights as for the single rectangle in **Table 1**. The increase in gain for the double rectangle is everywhere apparent in the table. However, the chief effect of the table is likely to be to make many folks wish they lived surrounded by very good earth. I did not have the heart to present the salt-water data.

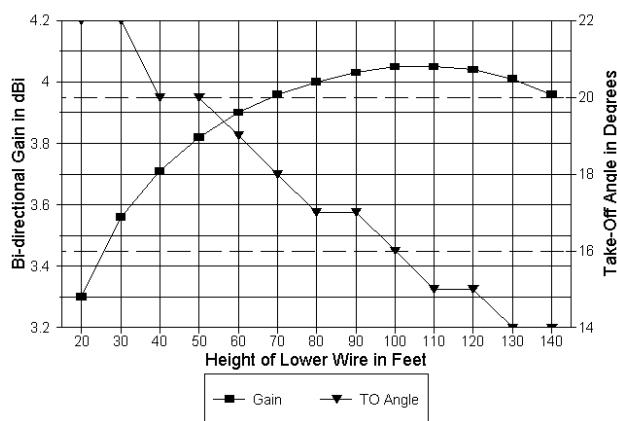
### The Result

Unlike the patterns for vertical mono-poles and dipoles, the SCV family of antennas display a highly bi-directional pattern. **Figure 11** presents the azimuth pattern for a double rectangle over average soil with a baseline height of 100 ft and a take-off angle of 16 degrees. The single rectangle presents a similar pattern. Note the clover-leaf in the center of the pattern: it represents remnant horizontally polarized radiation

**Table 2**  
160-Meter Double Rectangle Gain, TO Angle, and Feedpoint Impedance Over Various Soils, with Different Antenna Heights Above Ground

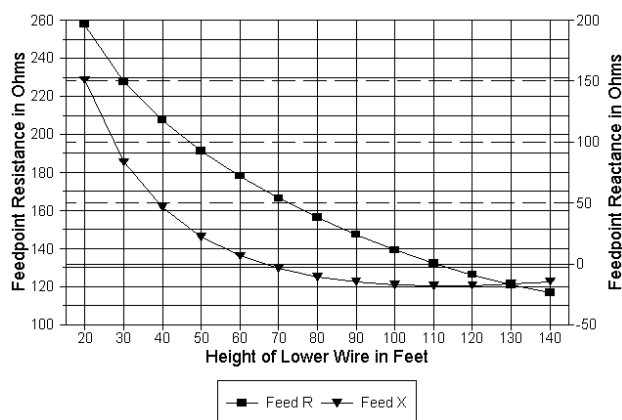
Antenna Height (in feet)	Soil Type	Gain (dbi)	Take-Off Angle (degrees)	Feed-point Impedance (R +/- jXΩ)
20-91	Very Poor Soil	0.54	27	274 + j120
	Poor Soil	2.07	24	264 + j139
	Average	3.30	22	258 + j151
	Very Good	5.46	16	237 + j156
60-131	Very Poor Soil	1.38	23	171 - j6
	Poor Soil	2.83	21	174 + j2
	Average	3.90	19	178 + j7
	Very Good	6.06	14	179 + j14
100-171	Very Poor Soil	1.60	20	131 - j18
	Poor Soil	3.01	18	136 - j17
	Average	4.05	16	139 - j17
	Very Good	6.50	12	143 - j15
<b>Soil types</b>	<b>Conductivity (S/m)</b>	<b>Dielectric Constant</b>		
Very Poor Soil	0.001	5		
Poor Soil	0.002	13		
Average	0.005	13		
Very Good	0.0303	20		

**Double Rectangle: Gain and TO Angle**  
206x71'; Average Soil; 1.85 MHz

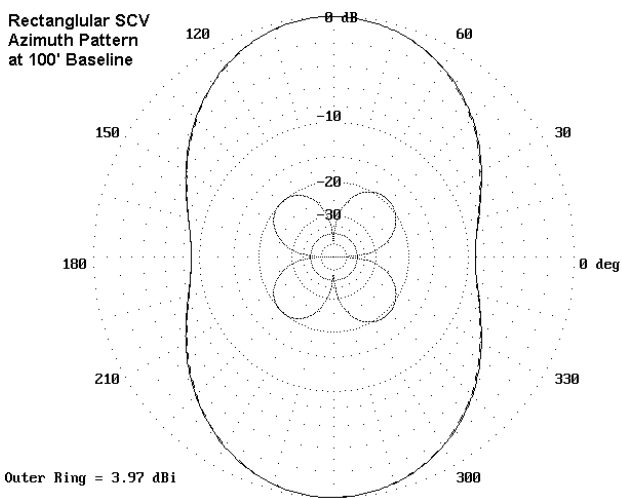


**Figure 9—Double 160-meter rectangle gain and take-off angle at various heights.**

**Double Rectangle: Feedpoint Impedance**  
206x70.5'; Average Soil; 1.85 MHz



**Figure 10—Double 160-meter rectangle feed-point impedance at various heights.**



**Figure 11—Azimuth pattern for the rectangle SCV, with horizontal and vertical components.**



which is not eliminable from any of the SCV configurations.

In general, deltas and square loops have more broadly oval patterns with less front-to-side rejection than the rectangle. The half-square pattern is similar to that of the rectangle with a slightly greater front-to-side ratio.

To increase gain further—and in the process double the front-to-side ratio—requires no more altitude, but double the linear space for the antenna. K4VX's open double rectangle with a common center wire or the familiar bobtail curtain provide about 1.5 dB added gain and over 20 dB front-to-side ratio, with comparable take-off angles to the single and double rectangles shown here.<sup>4</sup> They are certainly antennas worth investigation if one has a linear space well over 400 ft long.

These notes are based on computer models of the rectangle, which itself has already been proven in the field. Where computer modeling is at its best is in developing systematic guidance data, and that has been the aim of these notes. Computer models assume level terrain, but the individual contemplating an antenna such as these might well use terrain analysis software by N6BV or K6STI to adjust expectations for the particular antenna site and its environs. The more data we gather in advance, the more realistic will be our expectations for any antenna we might think about building. The 160-meter rectangle makes a good case-in-point.

### But What About 80 Meters?

In our haste to review the rectangle at 160 meters, we have bypassed data for 80 meters, where the antenna is certainly a candidate for SCV use. The ideal ratio of length to fed-side for a 3.6 MHz rectangle is about 3.6:1. #12 AWG copper wire models in free space yielded maximum gain with dimensions of about 110 ft long by 31 ft high, at a figure about 0.25 dB higher than the 160-meter model. However, the feed-point impedance was about 8 Ω lower. These latter two properties result from the narrower shape of the 80-meter rectangle.

With these variations in mind, you can anticipate the values for the 80-meter rectangle that appear in **Table 3**. Only the values for average soil are shown, since the values for other soils are proportional, using the 160-meter charts as a guide. The height of maximum gain is just about half that for 160 meters. However, over ground, the gain is not as high as the corresponding 160-meter rectangle at twice the height.

Only at low heights is it advisable to feed the 80-meter single rectangle directly with coax. Indeed, the single rectangle above 160 meters is probably a worse choice than its companion double rectangle. Like the 160-meter version of the double rectangle, the 80-meter antenna displays a slight gain over a single rectangle with relative insensitivity to the spacing of the two wires. The sample model placed the wires 1-ft apart, with the cross-over wires spaced about 0.5 ft apart. Also like the 160-meter version, the model maintained the same length, but increased the height over the single rectangle by a small amount, ending up with a total height of 32.3 ft for resonance in free space.

**Table 4** shows the modeled values over average soil at heights ranging from 10 to 90 ft (with resultant top-wire heights ranging from 42.3 to 122.3 ft). The added gain over the single rectangle is evident, as is the reduced gain relative to corresponding 160-meter double rectangles. Like the 160-meter models, the height of maximum gain for the double rectangle is slightly and perhaps insignificantly higher than for the single rectangle.

The feed-point impedance at low heights is likely to benefit from the use of parallel feeders and an antenna tuner. At higher levels, a  $\frac{1}{4} \lambda$  matching section of 75 Ω cable will provide a 50 Ω coax match over a small portion of the band. However, the large scale changes in reactance suggest that

**Table 3**

#### 3.6 MHz Single Rectangle: Properties Over Average Soil at Various Heights.

Soil Type	Baseline Height (ft)	Gain (dBi)	T-O Angle (degrees)	Feed Impedance R +/- jX Ω
Average	10	2.21	25	50 + j38
(C=0.005, DC=13)	20	2.76	23	40 + j15
	30	2.98	21	34 + j7
	40	3.08	20	30 + j3
	50	3.11*	18	27 + j2
	60	3.08	17	25 + j2
	70	3.01	16	23 + j2
	80	2.88	15	22 + j3
	90	2.69	14	21 + j4

Note 1. \* = Height of maximum gain

Note 2. Dimensions of single rectangle = 110 ft baseline length, 31 ft height.

Construction: #12 AWG copper wire.

**Table 4**

#### 3.6 MHz Double Rectangle: Properties Over Average Soil at Various Heights.

Soil Type	Baseline Height (ft)	Gain (dBi)	T-O Angle degrees	Feed Impedance R +/- jX Ω
Average	10	2.40	25	212 + j132
(C=0.005, DC=13)	20	2.98	23	164 + j39
	30	3.23	21	138 + j6
	40	3.36	19	120 - j8
	50	3.42	18	107 - j14
	60	3.43*	17	97 - j15
	70	3.37	15	90 - j13
	80	3.26	15	85 - j9
	90	3.08	14	82 - j5

Note 1. \* = Height of maximum gain

Note 2. Dimensions of double rectangle = 110 ft baseline length, 32.3 ft height.

Construction: #12 AWG copper wire.

parallel feeders and a tuner may be best for operation over the entire band.

This basic data, when combined with sensible adjustments to the 160-meter data, should provide reasonable guidance for our expectations should we decide to build one of these antennas. Before we make such a decision, we shall want to compare these data with those in the last episode on deltas. But let's not be too hasty. Final decisions should await a fuller story on the open-ended cousin to these two loops: the half-square.

### Notes

<sup>1</sup>For the "magnetic slot" and "double magnetic slot," see Russell E. Prack, K5RP, "Magnetic Radiators—Low Profile Paired Verticals for HF," *The ARRL Antenna Compendium*, Vol. 2 (Newington: ARRL, 1989), pp 39-41. However, the elongated loop or "oblong" and its relationship to the square quad has been well-known for a long time. See, for example, the reference to this subject in Karl Rothammel, Y21BK, *Antennenbuch* (Berlin: Militarverlag der DDM, 1984), p 230, where a ratio of about 2.4:1 is recommended for the vertically polarized version. Reference is made therein to work by G6LX.

<sup>2</sup>Brian Beezley, K6STI, "A Gain Antenna for 28 MHz," *QST* (July, 1994), p 70.

<sup>3</sup>In the HF region, we can use a simpler approximation:  $(2.8 + 1.4 \log F)$ , where F = the frequency in MHz for #12 copper wire. When expressed in terms of natural logarithms, R approaches the Fibonacci constant times  $\ln(100 F)$ .

<sup>4</sup>For the open double magnetic slot, see Lew Gordon, K4VX, "The Double Magnetic Slot Antenna for 80 Meters," *The ARRL Antenna Compendium*, Vol. 4 (Newington: ARRL, 1995), pp 18-21. ■